

# High Voltage Circuit Breakers Vibration Signal Analysis based on Power Spectrum Estimation



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**ABSTRACT:** Power spectrum estimation method refers to vibration or random signal power spectrum calculation method, The power spectrum analysis is an important part of the signal analysis, can be used to detect the periodic component and the percentage of the signal. And it has important implications for the diagnosis of mechanical failure.

**Keywords:** Spectrum Estimation, Vibration Signal, AR Model

**Received:** 25 May 2015, Revised 26 July 2015, Accepted 2 August 2015

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## 1. Introduction

Generally speaking, power spectrum estimation method can be divided into classical spectral estimation method and modern spectral estimation method. Classical spectral estimation methods are defined by the finite-length sequence to estimate directly, there are two ways to achieve: (1) Estimate the correlation function first, and then get the power spectrum estimation by the Fourier transform; (2) Link The power spectrum and the square of the signal amplitude-frequency characteristics. No matter the use of which one way, the common problem is the characteristic estimation of variance is not good, and the estimated value of ups and downs along the frequency axis is very fierce. With the longer of the data, the phenomenon will be more serious.

Classical spectral estimation method can be divided into direct and indirect methods, direct method is to use a fast Fourier algorithm (FFT) for a finite number of sample data to obtain a power spectrum of the Fourier transform method, also called periodogram; The indirect method is to get the autocorrelation function estimates of sample data first, then performs Fourier transform of the power spectrum. Since direct method to obtain the power spectrum estimation spectral presence some shortcomings such as undulating curves, spectral resolution is not high, several improved algorithms proposed, such as Bartlett Law and Welch method. Modern spectral estimation is mainly made for the classic low resolution spectral estimation variance and poor performance and other issues, from the way the modern spectral estimation method can be divided into parametric model method and non-parametric model method, power spectrum estimation based on parametric modeling is an important content of modern power spectrum estimation. The main parameters of the model method has AR model method, MA and ARMA model method model law.

## 2. Classic Spectrum Estimation

### 2.1 Periodogram Method

Classical periodogram method (Periodgram) also called direct method, first calculate the Fourier transform  $X(e^{j\omega})$  of the random signal's  $N$  observations  $x(n)$ , then take the square of the amplitude, and dividing by Sequence of samples' number  $N$  as the sequence  $x(n)$  estimates of the true power spectrum, namely:

$$\hat{p}_{per} = \frac{1}{N} |X(e^{j\omega})|^2 \quad (1)$$

When the sample sequence number  $N$  tends to infinity, the expected value of the direct method is equal to the power spectrum's real value. The spectral estimated resolution is proportional to  $\frac{2\pi}{N}$ , and the data truncation and windowing, making that when the spectral curve larger the more dramatic ups and downs.

## 2.2 Direct Method's Improved Algorithms

For the direct method to obtain the power spectrum estimation, when the data length is too large, the spectrum curves undulating; if the data length is too small, the spectral resolution is not high, So people also proposed several improved algorithms, such as Bartlett method and Welch method.

### 2.2.1 Bartlett Method

Bartlett average periodogram method is getting  $N$  points finite-length sequence  $x(n)$  segments seeking re average. Assuming that  $x(n)$  into  $L$  segments, and  $M$  samples per segment, Thus  $N = LM$ , paragraph  $i$  sample sequence can be written as:

$$x^i(n) = x(n + (i-1)M) \quad 0 \leq n \leq M-1, 1 \leq i \leq L \quad (3)$$

$$\hat{P}_{per}^i(\omega) = \frac{1}{M} \left| \sum_{n=0}^{M-1} x^i(n) e^{-j\omega n} \right|^2 \quad (4)$$

### 2) Welch Method

Welch method of Bartlett law was amended in two respects: First, select the appropriate window function  $\omega(n)$ , and directly adding in front of the cycle diagram calculation, so that the resulting cycle Pictured each segment:

$$\hat{P}_{per}^i(\omega) = \frac{1}{MU} \left| \sum_{n=0}^{M-1} x_n^i(n) \omega(n) e^{-j\omega n} \right|^2 \quad (5)$$

in the formula,

$$U = \frac{1}{M} \sum_{n=0}^{M-1} \omega^2 \quad (6)$$

Is normalization factor. Windowed advantage is that no matter what kind of window functions can make non-negative spectrum estimation; and, when segmented, can overlap between segments, this would reduce the variance.

Corresponding to the sum, and then averaged to obtain an average Is the power spectrum:

$$\bar{P}_{per} = \frac{1}{L} \sum_{i=1}^L P_{per}^i(\omega) \quad (7)$$

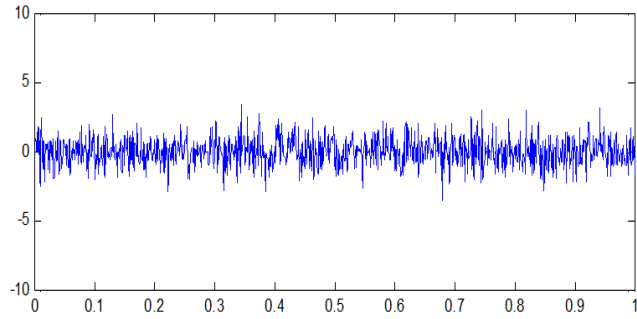
## 2.3 The indirect method

The indirect method estimates the correlation function by the sequence  $x(n)$  first, Then do Fourier transform of the autocorrelation function evaluation, we get the the power spectrum estimation of sequence,  $x(n)$  namely:

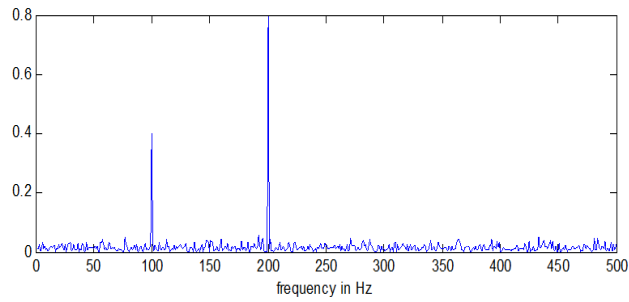
$$\hat{P}_{BT}(\omega) = \sum_{m=-N}^N \hat{R}_x(m) e^{-j\omega m} \quad (8)$$

## 2.4 Simulation Analysis

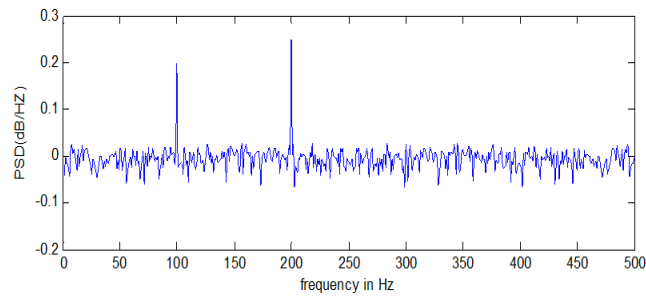
Take a random sequence of noisy  $x(n) = 4 \cos 100\pi n + 7 \cos 200\pi n + z(n)$ ,  $z(n)$  is a random white noise, Several methods were used at the power spectrum estimation, as well as to facilitate comparison, the power spectral density do normalization, as shown in Figure 1.



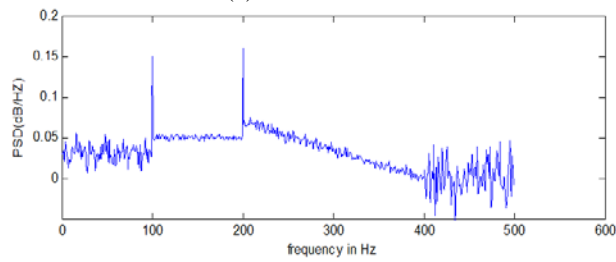
(a) Time Domain



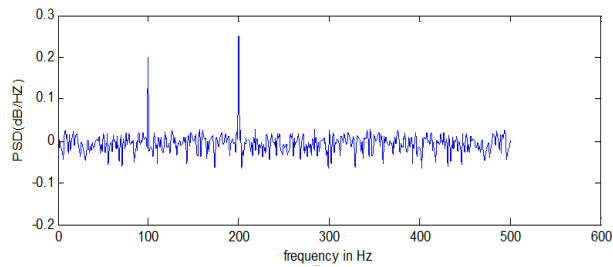
(b) FFT Spectrum Analysis



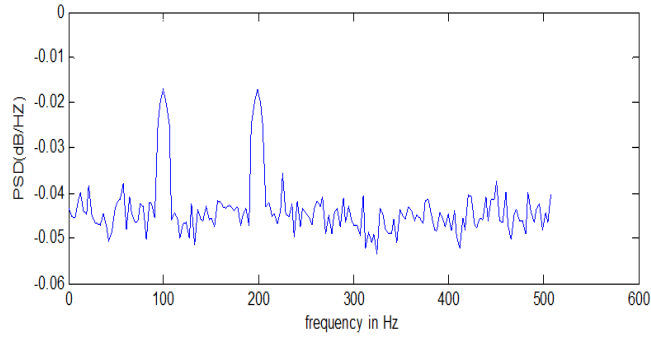
(c) Direct method



(d) Indirect method



(e) Bartlett method

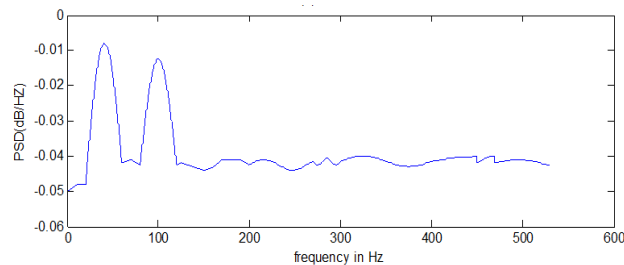


(f) Welch method

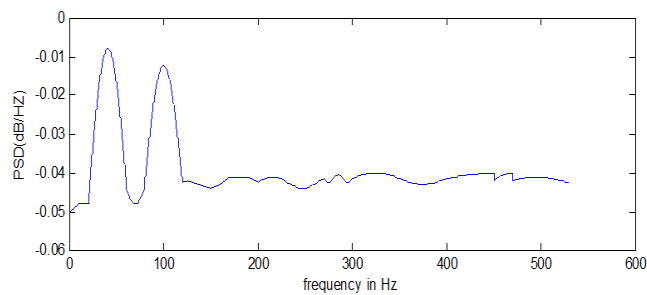
Figure 1. Comparative sequence  $x(n)$  of different power spectrum estimation

Although the Bartlett method does not specify window function, but for finite-length sequence, the equivalent rectangular windows; Meanwhile to facilitate comparison, Welch method also takes a rectangular window.

From the spectral estimation effect, the resolution of the indirect method is significantly better than the direct method; Comparison of Figure 1 (e), (f), we can see that Welch method's spectral estimation's error is small than method's. This is because although have adopted the rectangular window, But in the algorithm, when segments the same sequence, Welch method have overlap between the paragraphs, but Bartlett method didn't, Experiments show that Overlapping  $\frac{1}{2}$  data length is appropriate, therefore, the estimated variance characteristics, the former than the latter; If the window function better replaced Hamming window, Blackman window, using the Welch method was relatively smoother curve, as shown in Figure 1.



(a) Hamming window

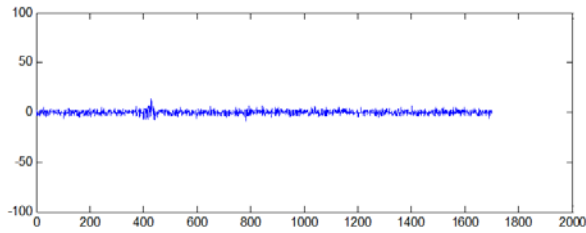


(b) Gabe Blackman window

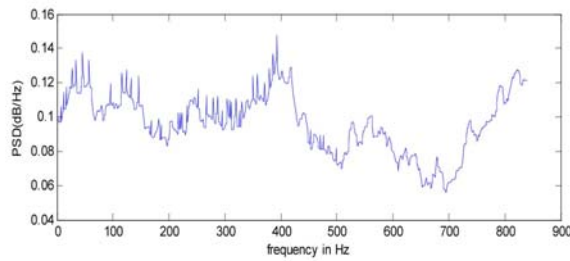
Figure 2. Welch power spectrum estimation method

Take a group of scene acquiring breaker vibration signal, using the indirect method, Welch method to estimate the power spectrum distribution respectively.

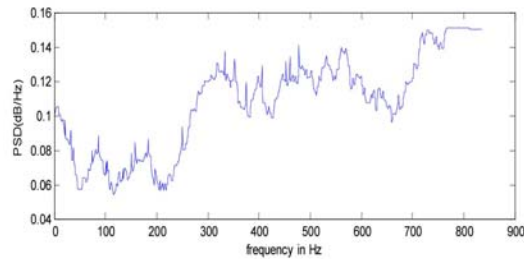
Despite the length of the vibration signal reaches  $N = 1500$ , can be counted as long data samples, Spectrum estimation results are still not satisfactory. Analysis found that although Welch method does a considerable improvement, but still essentially unknown outside the work area data assume that the data is zero, equivalent data windowing, resulting in reduced resolution and spectral estimation instability; The indirect method is the autocorrelation function of the sequence directly as Fourier transform, so the impact will be better than the Welch method.



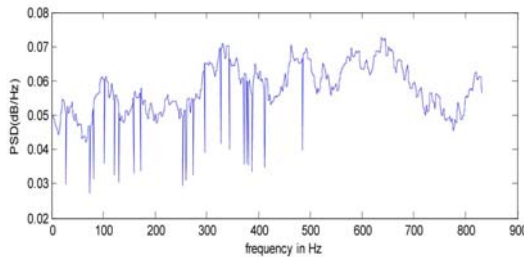
(a) The original signal



(b) FFT spectrum analysis



(c) The indirect method



(d) Welch Act (Gabe Blackman window)

Figure 3. Two power breaker vibration signal spectrum estimation

Modern spectral estimation methods are no longer simply the unknown data is assumed to be observed outside the region and to zero, but first observation data signals to estimate the model parameters, seeking the estimated signal power spectrum in accordance with the output power of a model, avoided the data hypothetical question of data observation outside the area,

spectrum estimation performance is superior to the classical spectrum estimation. Therefore, the following attempts to adopt modern spectral estimation methods.

### 3. AR model of the Power Spectrum Estimation

The power spectrum estimation based on the parametric modeling is an important part of modern power spectrum estimation, Its purpose is to improve the frequency resolution of the power spectrum estimation. Modern spectral estimation is parametric model spectrum estimation method based on random process, It can also be called a parametric model approach or simply model approach. Spectrum estimation method based on the general parameters of the model according to the following three steps:

- (1) To be determined or estimated random process to select a reasonable model;
- (2) Model parameter estimation based on known observation data;
- (3) Model parameter estimation was used to calculate the power spectrum.

Modern spectral estimation model is divided into AR model, MA model and the ARMA model. model, also called moving average model, is the all-zero model; model is also known as auto-regressive moving average model, the “pole-zero” models. or course of any process can be represented by an infinite order, Even if you choose an appropriate model spectrum estimation procedure, you can always possible to select a sufficiently high order of approximation to the stochastic process modeled. Since accurate estimation of model parameters can be obtained by solving a set of linear equations, and for and model of the power spectrum estimation, accurate estimation of its parameters need to solve a set of higher-order nonlinear equations. Practice shows, AR model has a power spectrum for short data analysis, calculation is simple, get a smooth spectral resolution advantages, therefore model has been widely used.

#### 3.1 AR Model of the Power Spectrum Estimation Fundamentals

A number of stationary random processes can be generated by using white noise excitation linear time-invariant systems. Figure 4,  $x(n)$  is composed of an input sequence  $u(n)$  excitation output of a causal linear shift invariant discrete-time system  $H(n)$ .

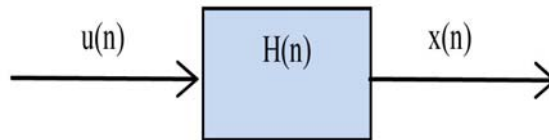


Figure 4. Parameters of the system model

If the discrete stationary random sequence  $x(n)$  can be linear differential equation (8) to describe:

$$x(n) + \sum_{i=1}^p a_i x(n-i) = u(n) + \sum_{k=0}^q b_k e(n-k) \quad (8)$$

Where  $u(n)$  is a discrete white noise, called  $\{x(n)\}$  is ARMA process.

Formula (8) on both sides were taken Z transform, and assuming  $b_0 = 1$ , can be obtained by formula (9), that is

$$H(z) = \frac{B(z)}{A(z)} = \frac{\sum_{k=0}^q b_k z^{-k}}{\sum_{k=0}^p a_k z^{-k}} \quad (9)$$

of formula (9)

$$A(z) = 1 + \sum_{k=1}^p a^k z^{-k} \quad (10)$$

$$B(z) = 1 + \sum_{k=1}^q b^k z^{-k} \quad (11)$$

Thereby obtaining the output sequence is a power spectrum

$$p_x(e^{-j\omega}) = \frac{\sigma^2 |B(e^{-j\omega})|^2}{|A(e^{-j\omega})|^2} \quad (12)$$

If  $b_1, b_2, \dots, b_q$  are all zero, the formula (8) becomes:

$$x(n) = -\sum_{k=1}^p a_k x(n-k) + u(n) \quad (13)$$

This model is called autoregressive model, referred to as *AR*, model. Where  $u(n)$  is a white noise with variance  $\sigma^2$  sequence,  $p$  is the order of the *AR* model,  $a_k$  the *AR* model parameters,  $k = 1, 2, \dots, p$ . For the *AR* model, signal  $x(n)$  from the previous value itself several excitation current value and generating a linear combination, and the accurate estimation of *AR* model parameters can be obtained by solving a set of linear equations.

Also, the formula (12) becomes:

$$P_x(e^{-j\omega}) = \frac{\sigma^2}{\left| 1 + \sum_{k=1}^p a_k e^{-j\omega k} \right|^2} \quad (14)$$

That random signal  $x(n)$  of the *AR* model of the power spectrum. It can be seen, to power spectrum estimation, we must obtain the *AR* model order  $p$ , model parameters  $a_1, a_2, \dots, a_p$  and  $\sigma^2$ .

### 3.2 Choose the AR Model Order

Select the appropriate *AR* model order *AR* spectral estimation is a key issue, Order  $p$  *AR* model is generally not known in advance, you need a slightly larger pre-selected value, determined in the recursive process. When using the Levinson recursive algorithm (Levinson-Durbin), can be given by the low-level to high-end of each set of parameters, and the minimum prediction error model of the power  $P_{min}$  (equivalent to white noise sequence variance  $\sigma^2$  is decreasing. Intuitively,  $p$  when the prediction error reaches a specified desired value, or no longer change, then the correct order of the order that is to be selected.

Because prediction error power  $p$  is a monotonically decreasing, This value is reduced to just the right number is often hard to choices, Order selection too low will have a greater bias, and may cause peaks can't tell; Select the parameters to be estimated too high is because too much performance degradation caused by spectral estimation variance. To this end, there are several different criteria have been proposed, the following are a few of the more common criteria.

#### 1) Several Common Criteria to Determine the AR Model Order

##### a) Final Prediction Error Criterion (FPE)

Make  $FPE(r)$  for the application of the minimum value of the order of  $r$

$$FPE(r) = P_r \frac{N+r+1}{N-r-1} \quad (15)$$

Where  $r$  is the order of the prediction error  $P_r$  power,  $N$  is the number of samples, according to the following formula to calculate  $P_r$ .

$$P_r = \sum_{j=0}^r a_j R(j) \quad (16)$$

$a_j$  as a model parameter,  $R(j)$  of the signal  $x(n)$  of the autocorrelation function.

**b) Information Criterion (AIC)**

Make  $AIC(r)$  for the application of the minimum value of the order of  $r$

$$AIC(r) = N \ln(P_r) + 2r \tag{17}$$

Wherein the parameters  $N, P_r$  same as defined above. When the order  $r$  of a start to increase,  $FPE(r)$  and the  $AIC(r)$  are obtained at a certain minimum value of the  $r, r$  at this time is the most appropriate as the order  $p$ .

When the data is shorter,  $FPE, AIC$  given the low number of bands, and the results are given in the two is basically the same. When the sample size  $N$  tends to infinity, the amount of information  $FPE(r)$  and  $AIC(r)$  equivalent, when  $N \rightarrow \infty$ , criteria for selecting the correct order of the probability of error is not zero.

**c) The Minimum Description Length (MDL)**

Rissanen proposed an alternative information criterion -minimum description length ( $MDL$ ), criterion is based on the process to make the model corresponding probability distribution probability approaching the maximum possible distribution of the actual process of thinking. It is defined as the amount of information.

$$MDL(r) = N \ln(P_r) + r \ln N \tag{18}$$

$MDL$  statistical information criterion is consistent, experimental results show that the short data,  $AR$  order should be selected in the  $\frac{N}{3} \sim \frac{N}{2}$  range will have good results.

Clayton Ulrych and proved that for a short segment,  $FPE, AIC$  work was not good.

**d) The Empirical Formula**

Engineering frequently using the following formula:

$$P = \frac{2^{1-k} N}{\ln(2^{1-k} N)} \tag{19}$$

Where  $N$  is the length of the data signal,  $k$  is the wavelet packet decomposition level.

According to the formula shows that order of a), b), c) three criteria estimates on the prediction error power  $P_r$ , by the formula (16),  $P_r$  consists  $x(n)$  of a sequence of the autocorrelation function  $r(j)$ ,  $AR$  model parameters two parts; According to the definition of both, when the sequence  $x(n)$  changes in the autocorrelation function  $r(j)$ ,  $AR$  model parameters  $a(j)$  values also changed; The empirical formula is defined by commonly known, order  $p$  by the sequence length  $N$ , the number of wavelet packet  $k$  decisions and sequence values  $N$ , the noise itself is irrelevant; so the four methods, in addition to commonly used empirical formula, the order  $p$  under the other three criteria were randomly change (see Table 1), Wherein  $i$  is a random number of operations,  $i = 4$ .

It should be noted that the above is only a few criteria for order selection provides a basis, for the study to a specific signal, and ultimately choose their  $AR$  model order, you need to practice to compare the results of multiple.

**3.3 Simulation Analysis**

Take a sequence of noisy  $x(n) = 3 \cos(200\pi n) + 5 \cos(100\pi n) + e(n)$ ,  $e(n)$  is a random Gaussian white noise, sampling rate  $f_s = 100\text{Hz}$ , take  $N = 1024$  points, four methods are solved on  $AR$  model parameters  $p$ , the results in Table 1.



As can be seen by the sequence  $x(n)$  of the expression, because  $e(n)$  is a random Gaussian white noise, random values generated by each operation, so that the sequence  $x(n)$  values have changed.

$i=1,2,3,4$	AIC Criterion	FPE Criterion	MDL Criterion	Empirical formula
Pi	42	42	26	26
Pi	41	41	21	26
Pi	50	50	21	26
Pi	42	42	20	26

Table 1. Random distribution of each criterion with random Gaussian white noise signal number order

Take another noisy sequence  $x(n) = 3 \sin(400\pi n) + 5 \cos(402\pi n) + x_e(n)$ ,  $x_e(n)$  is a noisy square wave,  $x_e(n)$  is not randomly distributed, In the same bandwidth (assuming here are  $[0,1]$ ) of the premise, change the sample rate  $f_s$  (equivalent to changing the sequence length  $N$ ), the results in Table 2.

N	AIC Criterion	FPE Criterion	MDL Criterion	Empirical formula
P(N=2048)	77	77	11	46
P(N=1024)	50	50	18	26
P(N=512)	28	28	9	15
P(N=256)	17	17	7	9
P(N=128)	8	8	8	6
P(N=64)	8	8	1	4
P(N=32)	5	5	1	3

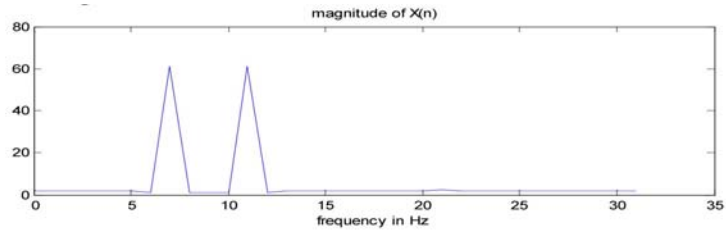
Table 2. The following four kinds of sequences of different lengths in order to determine

Found from Table 2, the sequence  $x(n)$  by a longer data ( $N = 2048$ ) becomes short data ( $N = 32$ ) of the process, the method of determining the order of four substantially decreased. Detailed analysis of the following variation of the order under *AIC*, *FPE*, *MDL* criterion  $p$ 's:

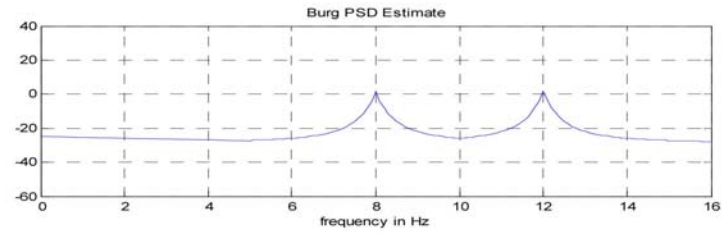
(1) *AIC* criteria and guidelines for the *FPE*, regardless of size  $N$  samples, the results given both are the same, when the number of samples  $N$  is shorter, the order declined rapidly, but compared with the other two methods, the value is relatively high;

(2) For *MDL* criterion when short data is much smaller than the estimated value given by  $\frac{N}{3}$   $\frac{N}{2}$  range, the power spectrum estimation done, the frequency resolution is very poor, or even not affect the resolution of frequency components; For short data, *AR* order should be selected in the  $\frac{N}{3}$   $\frac{N}{2}$  range will have good results. I try several randomly noisy sequence analysis results of many experiments show that when changes in the order  $r$   $\frac{N}{3}$   $\frac{N}{2}$  range, the result is the estimation of each  $[N/3]$ .

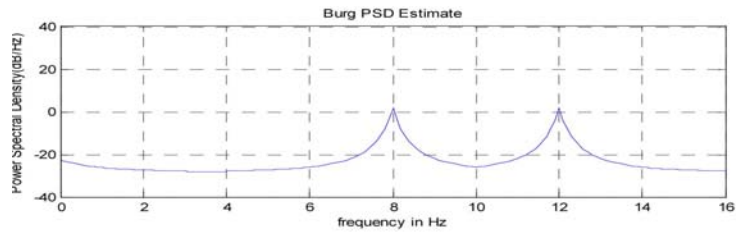
Take the following sequence of noisy  $x(n) = 3 \cos(2\pi \times f_1 \times n) + 5 \cos(2\pi \times f_2 \times n) + 0.1e(n)$ ,  $f_1=8\text{Hz}$ ,  $f_2=12\text{Hz}$ ,  $e(n)$  is a Gaussian white noise, four methods (*AIC*, *FPE*, *MDL*, commonly used empirical formula) results are given as follows:  $P_1=P_2=6$ ,  $P_3=5$ ,  $P_4=3$ ; Here Conclusion Based on the above (2) said, adding a case under that criteria when short data, order  $P_3$  take  $[N/3]$ ; In determining the order of the *AR* model  $r$  premise Burg method using the estimated power spectrum, spectral estimation results shown in Figure 5.



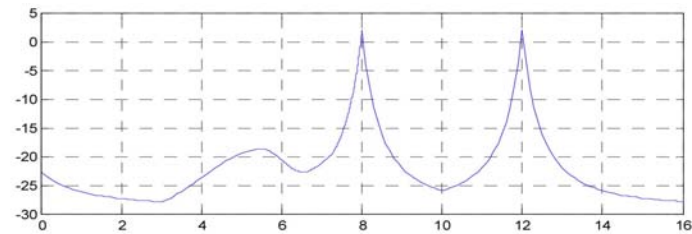
(a) FFT spectrum analysis



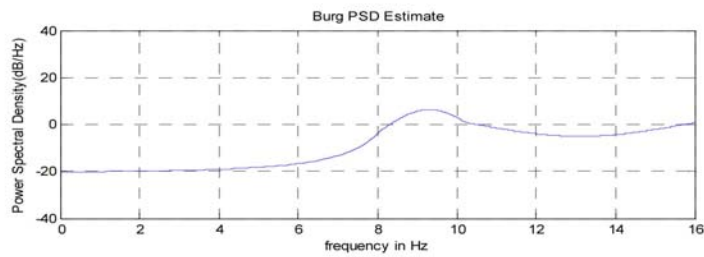
(b)  $P_1 = P_2 = 6$



(c)  $P_3 = 5$



(d)  $P_3 = 10$



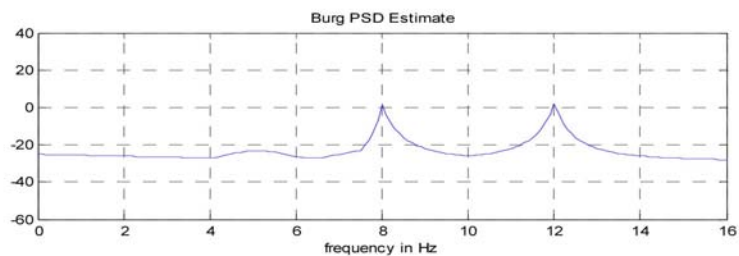
(e)  $P_4 = 3$

Figure 5. Power spectral analysis method under different criteria Burg

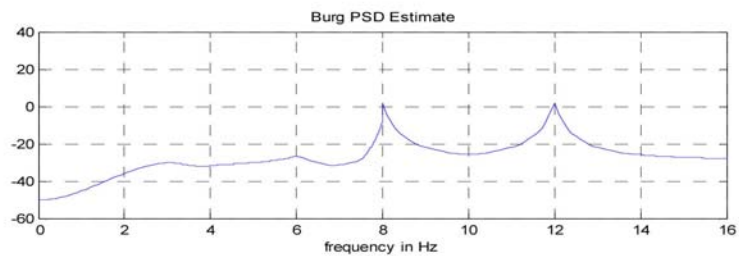
When the data sequence is short, that is  $N = 32$ , the order of  $AIC$ ,  $FPE$  standards and guidelines for obtaining similar  $MDL$ , when the order of  $P_1, P_2, P_3$  were taken Figure 5 (b), (c) the power spectrum of the frequency curve can be clearly distinguished  $f_1 = 8Hz$ ,  $f_2 = 12Hz$  two peaks, this is consistent with the results of (a) the fast Fourier algorithm (FFT) spectrum analysis; the results of empirical formula used  $P_4 = 3$ , the spectral curve is too smooth, the spectral resolution is poor, unable to distinguish  $f_1 = 8Hz$ ,  $f_2 = 12Hz$  two peaks (Figure 5 (e)), this is because we can see by the empirical formula, the results of the order  $P_4$  only by the signal length  $N$ , wavelet packet decomposition level  $k$  to decide, regardless of the frequency components of the signal; Figure 5 (d)  $MDL$  criterion is also used, the order taking  $P_3' \lfloor \frac{N}{3} \rfloor = 10$ , the effect of the power spectrum estimate (b), (c), as can clearly distinguish the two frequency peaks; Note that, in order to continue to increase during  $P_3' \frac{N}{2}$ , the spectrum smoothing section undulating curve intensified, new peaks appeared, the spectrum changes shown in Figure 6.

When the order  $r$  to 13, starting at  $f \approx 4Hz$  spectrum spurious peaks (see Figure 6 (c)); continue to increase  $r$ , the spectrum has been found in more false peaks, figure 6 (d), (e), (f) shown below, but the pseudo-peak energy relative to the two main peaks appear ( $f = 8Hz$ ,  $f = 12Hz$  Department) is very weak, from the curve was still able to clearly distinguish the two peaks.

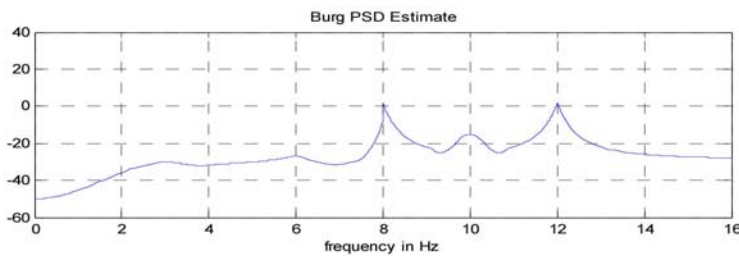
Therefore, when a shorter data using  $MDL$  criterion, the order can be directly taken  $\lfloor \frac{N}{3} \rfloor$ , also available at  $(\frac{N}{3})$ , within the range of  $\frac{N}{2}$  values, spectral estimation results are basically unaffected.



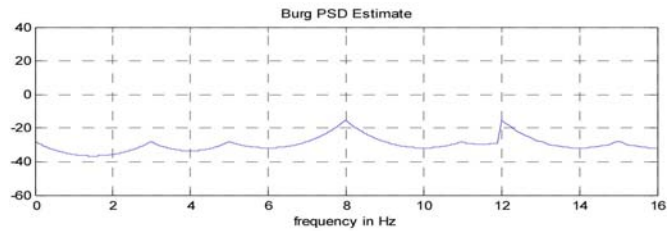
(a)  $r=11$



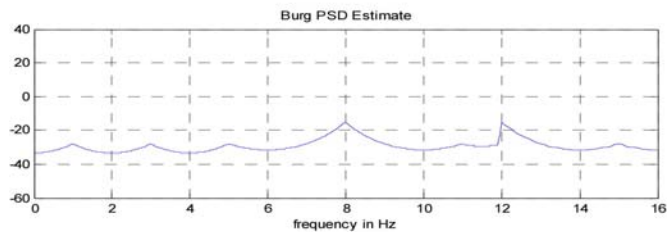
(b)  $r=12$



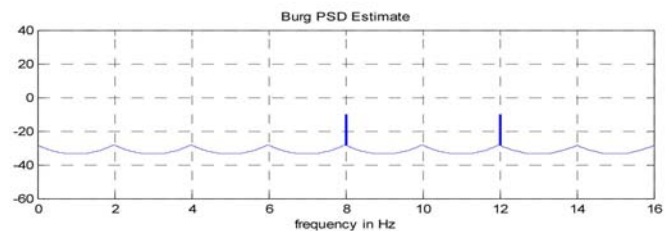
(c)  $r=13$



(d)  $r=14$



(e)  $r=15$



(f)  $r=16$

Figure 6. Different orders of power spectrum estimation

#### 4. Conclusion

Based on the above analysis, although some guidelines have been proposed, but no single criterion can be well suited to a variety of situations, several selection criteria described in the order of only provide a basis for determining the order of the final signal is often affected by the specific nature of the study, the analysis of the results of the actual need for several experiments.

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